ELEC 343, Home Work Assignment # 1:

Follow through Example 1A, and do Study Problem, SP1.2-2, SP1.2-4 Follow through Example 1B, and do Study Problems SP1.3-1, SP1.3-3 Follow through Example 1D, and do Study Problems SP1.5-2 and SP1.5-3. Textbook Chapter Problem(s): 3, 8, 9, 15, and 16. Solutions:

$$SP1-2-2$$
From $SP1.2-1$, we have phasons of voltage and current:
$$\tilde{V} = 120^{\circ}, \ \tilde{I} = 12180^{\circ}$$
Then we have $\omega = 2.70.60 \approx 377 \text{ r/s}$

$$V = \sqrt{2} \cdot \cos(\omega t) = \sqrt{2} \cdot \cos(377.t)$$

$$I = \sqrt{2} \cdot \cos(\omega t + \pi) = \sqrt{2} \cdot \cos(377.t + \pi)$$

$$P = V \cdot I = \sqrt{2} \cdot \cos(377.t) \cdot \sqrt{2} \cdot \cos(377.t + \pi)$$

$$= 2\left(\frac{1}{2} \cdot \cos(754.t + \pi) + \frac{1}{2} \cos(-\pi)\right) =$$

$$= -1 + \cos(754.t + \pi)$$

SP 1.2-4 For simplicity, and without lack of generality, neglect resistance and assume the voltage is along the real axis. In this case the relationship between voltage and current is given by:

$$V \angle 0 = j(X_L - X_C)\tilde{I} + \tilde{E}$$

One can observe that the magnitude and phase of \tilde{E} controls the magnitude and direction of \tilde{I} . For example, if $\tilde{E}=2\angle 0^\circ$ and $X_L>X_c$, \tilde{I} will lead \tilde{V} . In contrast, if $\tilde{E}=0.5\angle 0^\circ$ and $X_L>X_c$, \tilde{I} will lag \tilde{V} .

$$\begin{split} \mathrm{SP1.3-1.} \quad \mathrm{Ni} &= \mathcal{R}_{\mathrm{ab}} \left(\Phi_1 + \Phi_2 \right) + \mathcal{R}_{\mathrm{bcda}} \, \Phi_1 \\ \Phi_1 &= \frac{1}{\mathcal{R}_{\mathrm{bcda}}} \left[\mathrm{Ni} - \mathcal{R}_{\mathrm{ab}} \left(\Phi_1 + \Phi_2 \right) \right] \\ \Phi_1 &= \frac{1}{358,099} \left[- (109,419)(2.547 \times 10^{-3}) \right] = 2.014 \times 10^{-3} \; \mathrm{Wb} \end{split}$$

SP1.3-3. With windings as shown in Fig. 1B-1 and with the center leg removed, the total mmf is

$$mmf_t = mmf_1 + mmf_2 = N_1 I_1 + N_2 I_2$$

= $(150)(9) + (90)(-15) = 0$

SP1.5-2. During steady-state conditions, the time rate-of-change of i_1 is zero; therefore, a voltage is not induced in the 2-winding. Hence, for the 2-winding open or short circuited $I_1 = \frac{V}{r_1} = \frac{12}{6} = 2$ A and $I_2 = 0$.

$$\begin{split} \mathbf{SP1.5-3.} \quad Z &= r_1 \, + \mathbf{j} \; \omega_e \, (L_{l1} \, + L_{m1}) = 6 \, + \mathbf{j} \, (100) (13.5 \, + 263.9) \times 10^{-3} \\ &= 6 \, + \mathbf{j} \, 27.74 = 28.38 \, \underline{/77.8^\circ} \\ \tilde{I}_1 &= \frac{\tilde{V}_1}{Z} = \frac{10/0^\circ}{28.38/77.8^\circ} = 0.352 \, \underline{/-77.8^\circ} \; A \end{split}$$

3. The reluctance of the iron is

$$\begin{split} \mathcal{R}_{\rm m} &= \frac{1}{\mu_{\rm i}\,\rm A} = \frac{(4)(0.25)}{(4000)(4\pi\times10^{-7})(0.05)^2} = 79,577\,\,\rm H^{-1} \\ L_{12} &= \frac{N_1\,N_2}{\mathcal{R}_{\rm m}} = \frac{(50)(100)}{79,577} = 0.0628\,\,\rm H \\ \\ L_{\rm m1} &= \frac{N_1^2}{\mathcal{R}_{\rm m}} = \frac{50^2}{79,577} = 0.0314\,\,\rm H \\ \\ L_{\rm m2} &= \frac{N_2^2}{\mathcal{R}_{\rm m}} = \frac{100^2}{79,577} = 0.1257\,\,\rm H \end{split}$$

8.
$$\tilde{V}_1 = \frac{10}{2\sqrt{2}} / 0^{\circ} = 3.54 / 0^{\circ} V$$

$$Z = r_1 + r_2' + j \omega_e (L_{11} + L_{12}') = 10 + 10 + j (2\pi)(30)(30 + 30) \times 10^{-3}$$

$$= 20 + j 11.31 \Omega$$

$$\tilde{I}_1 = \frac{\tilde{V}_1}{Z} = \frac{3.54 / 0^{\circ}}{20 + \text{j } 11.31} = 0.154 / -29.5^{\circ} \text{ A}$$

9. Since $\omega_e = 400$, $X_{m1} = 400 \Omega$. Neglecting the magnetizing current $i_1 = -i_2'$.

(a)
$$\tilde{V}_1 = \frac{2}{\sqrt{2}} / 0^{\circ} = \sqrt{2} / 0^{\circ}$$

$$\tilde{I}_1 = \frac{\tilde{V}_1}{(r_1 + r_2 + R_L) + j \omega_e (L_{|1} + L_{|2})} = \frac{\sqrt{2} / 0^{\circ}}{4 + j (400)(0.02)} = 0.158 / -63.4^{\circ} A$$
(b) $I_1 = \sqrt{2} 0.158 \cos(400t - 63.4^{\circ})$

$$\begin{split} \text{15.} \quad L_m(x) &= \frac{k}{k_o + x} \\ k &= \frac{N^2 \, \mu_0 \, A_i}{2} = \frac{(500)^2 (4\pi \times 10^{-7})(4 \times 10^{-4})}{2} = 2\pi \times 10^{-5} \\ k_0 &= \frac{l_i}{2 \mu_{ri}} = \frac{20 \times 10^{-2}}{(2)(1000)} = 10^{-4} \\ L_m(x) &= \frac{2\pi \times 10^{-5}}{10^{-4} + x} \, \text{H} \end{split}$$

The approximation for x > 0 is

$$L_{\rm m}(x) = \frac{2\pi \times 10^{-5}}{x} H$$

Now to find minimum value of x

$$\frac{2\pi \times 10^{-5}}{x} = 1.1 \frac{2\pi \times 10^{-5}}{10^{-4} + x}$$

Solving for x yields x = 1 mm. Thus, the approximate expression is 10% in error at x = 1 mm and less than this for x > 1 mm.

$$\begin{aligned} 16. \quad & L_m(x) = \frac{k}{k_0 + x} \\ & \frac{\partial L_m(x)}{\partial x} = \frac{-k}{(k_0 + x)^2} \\ & v = ri + (L_l + \frac{k}{k_0 + x}) \frac{di}{dt} - i \frac{k}{(k_0 + x)^2} \frac{dx}{dt} \\ & v = r\sqrt{2} \, I_s \cos \omega_e t - (L_l + \frac{k}{k_0 + t}) \, \omega_e \sqrt{2} \, I_s \sin \omega_e t - \sqrt{2} \, I_s \cos \omega_e t \left[\frac{k}{(k_0 + t)^2} \right] \end{aligned}$$
 Gathering terms

 $v = \left[r - \frac{k}{(k_0 + t)^2}\right] \sqrt{2} I_s \cos \omega_e t - \left(L_l + \frac{k}{k_0 + t}\right) \omega_e \sqrt{2} I_s \sin \omega_e t$

Taking the limit as $t \to \infty$

$$\mathbf{v} = \mathbf{r}\,\sqrt{2}\;\mathbf{I_s}\cos\omega_{\mathrm{e}}\mathbf{t} - \mathbf{L_l}\,\omega_{\mathrm{e}}\,\sqrt{2}\;\mathbf{I_s}\sin\omega_{\mathrm{e}}\mathbf{t}$$

which is the voltage equation for a linear r-L circuit. In phasor form,

$$\tilde{V} = (r + j \omega_e L_l)\tilde{I}$$